## General Description

The AAT4295/97 SmartSwitch is a member of Skyworks' Application Specific Power MOSFET (ASPM ${ }^{\text {TM }}$ ) product family. The AAT4295/97 is comprised of three/six low-side N -channel MOSFET switches that gate an applied load to ground. This device is intended for controlling RGB fashion lighting in portable products; it can also be used for a combination of general purposes where a load requires a low-side switch connection to ground. The AAT4295/97 simplifies design and layout limitations by eliminating the need for multiple GPIO control lines and discrete MOSFETs to control product features.

The state of each output channel is controlled with a single GPIO line via the EN/SET pin using Skyworks' Simple Serial Control ( $\mathrm{S}^{2} \mathrm{C}$ wire ${ }^{\mathrm{TM}}$ ) interface. After a short set of data pulses is sent to the EN/SET input and the line is pulled to logic high, the device is enabled with the desired switch configuration. In the enabled state, the AAT4295/97 typically consumes less than $3 \mu \mathrm{~A}$ of quiescent current. When EN/SET is pulled to a logic low, the device is disabled and each output switch is placed in a high impedance open state.

The AAT4295 and AAT4297 operate over an input voltage range of 1.8 V to 5.5 V , making them ideal for bat-tery-powered applications. The three-switch AAT4295 is offered in a Pb-free, 8-pin SC70JW package, while the six-switch AAT4297 is offered in a Pb-free, 12-pin TSOPJW package. Both devices are rated over the $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ temperature range.

## Features

- Input Voltage Range: 1.8 V to 5.5 V
- Independent Low-Side N -Channel MOSFET Switches:
- AAT4295: Three Channels
- AAT4297: Six Channels
- User-Programmable $\mathrm{S}^{2}$ Cwire Interface
- Single GPIO Controls State of Each MOSFET
- Low Quiescent Current: 3 3 A Typical
- Temperature Range: $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
- No External Components Required
- Available in Pb-Free Packages
- AAT4295 in 8-Pin SC70JW
- AAT4297 in 12-Pin TSOPJW


## Applications

- Cell Phones
- Multiple Low Power Switching
- Personal Communication Devices
- Portable Electronic Devices


## Typical Application



## Pin Descriptions

| Pin Number |  |  |  |
| :---: | :---: | :--- | :--- |
| AAT4295 | AAT4297 | Symbol | Function |
| 1 | 8 | VCC | Input supply voltage. |
| 2,3 | $9,10,12$ | N/C | No connection. |
| 4 | 11 | EN/SET | Input control pin using S2Cwire serial interface. The device records rising edges of <br> the clock and decodes them into eight states, which control the ON/OFF states of <br> the MOSFETs. See Table 1 for output settings. In addition, a logic low forces the <br> device into shutdown mode, reducing the supply current to less than $1 \mu A . ~ T h i s ~ p i n ~$ |
| should not be left floating. |  |  |  |

## Pin Configuration

AAT4295
SC70JW-8
(Top View)


AAT4297
TSOPJW-12
(Top View)


## Absolute Maximum Ratings ${ }^{1}$

| Symbol | Description | Value | Units |
| :---: | :--- | :---: | :---: |
| $V_{\text {cc }}$ to GND | Power Supply to GND | 6.0 | V |
| INx to GND | All Input (Drain) to GND | -0.3 to 6.0 | V |
| EN/SET | EN/SET to GND | -0.3 to 6.0 | V |
| $\mathrm{~T}_{\mathrm{J}}$ | Operating Junction Temperature Range | -40 to 150 | ${ }^{\circ} \mathrm{C}$ |

Thermal Information ${ }^{2}$

| Symbol | Description |  | Value | Units |
| :---: | :---: | :---: | :---: | :---: |
| $\theta_{\text {JA }}$ | Thermal Resistance | SC70JW | 225 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  |  | TSOPJW | 160 |  |
| $\mathrm{P}_{\mathrm{D}}$ | Maximum Power Dissipation | SC70JW | $440{ }^{3}$ | mW |
|  |  | TSOPJW | $625^{4}$ |  |

[^0]
## Electrical Characteristics ${ }^{1}$

$\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} ; \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$, unless otherwise noted. Typical values are $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| Symbol | Description | Conditions | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {cc }}$ | Supply Voltage Range |  | 1.8 |  | 5.5 | V |
| $\mathrm{I}_{\mathrm{Q}}$ | Quiescent Current | $\mathrm{V}_{\mathrm{cc}}=5 \mathrm{~V}, \mathrm{EN} / \mathrm{SET}=\mathrm{V}_{\mathrm{cc}}, \mathrm{I}_{\mathrm{OUT}}=\mathrm{No}$ Load, All Switches On |  | 3.0 | 10 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {Q(OFF) }}$ | Off Supply Current | EN/SET $=0, \mathrm{~V}_{\text {cC }}=5 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}$ Open |  |  | 1.0 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {DS(OFF) }}$ | Off Switch Current for Any Switch | $\mathrm{EN} / \mathrm{SET}=0, \mathrm{~V}_{\text {CC }}=5 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=0$ |  | 0.1 | 1.0 | $\mu \mathrm{A}$ |
| $\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}$ | On-Resistance | $\mathrm{V}_{\text {cc }}=5 \mathrm{~V}$ |  | 1.9 | 6.0 | $\Omega$ |
|  |  | $\mathrm{V}_{\mathrm{cc}}=3.6 \mathrm{~V}$ |  | 2.1 | 7.0 |  |
| $\mathrm{T}_{\text {CRDS }}$ | On-Resistance Temperature Coefficient |  |  | 2800 |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
| Ton | Output Turn-On Time ${ }^{\text {² }}$ | $\mathrm{V}_{\text {IN }}=5 \mathrm{~V}, \mathrm{R}_{\text {PULLUP }}=250 \Omega, \mathrm{C}_{\text {OUT }}=0.1 \mu \mathrm{~F}$ |  | 0.6 | 2.7 | $\mu \mathrm{s}$ |
| EN/SET |  |  |  |  |  |  |
| $\mathrm{V}_{\text {EN(L) }}$ | Enable Threshold Low | $\mathrm{V}_{\text {IN }}=1.8 \mathrm{~V}$ |  |  | 0.3 | V |
| $\mathrm{V}_{\text {EN(H) }}$ | Enable Threshold High | $\mathrm{V}_{\text {IN }}=5.5 \mathrm{~V}$ | 1.4 |  |  | V |
| $\mathrm{T}_{\text {EN/Set Lo }}$ | EN/SET Low Time |  | 0.5 |  | 75 | $\mu \mathrm{s}$ |
| $\mathrm{T}_{\text {EN/SET_HI_min }}$ | Minimum EN/SET High Time |  |  | 50 |  | ns |
| $\mathrm{T}_{\text {EN/SET_HI_MAX }}$ | Maximum EN/SET High Time |  |  |  | 75 | $\mu \mathrm{s}$ |
| TofF | EN/SET Off Timeout |  |  |  | 500 | $\mu \mathrm{s}$ |
| $\mathrm{T}_{\text {Lat }}$ | EN/SET Latch Timeout |  |  |  | 500 | $\mu \mathrm{s}$ |
| $\mathrm{I}_{\text {EN/SET }}$ | EN/SET Input Leakage |  | -1 |  | 1 | $\mu \mathrm{A}$ |

[^1]
## Typical Characteristics

Quiescent Current vs. Input Voltage


Off-Supply Current vs. Temperature

$\mathrm{R}_{\mathrm{DS}\left(\mathrm{ON}_{\mathrm{N}}\right)} \mathrm{vs}$. Temperature
$\left(V_{\text {IN }}=3.6 \mathrm{~V} ; \mathrm{I}_{\text {LOAD }}=20 \mathrm{~mA}\right)$


Quiescent Current vs. Temperature


$$
\mathrm{R}_{\mathrm{DS}(\mathrm{ON})} \text { vs. Input Voltage }
$$




## Typical Characteristics

EN/SET Timeout vs. Input Voltage


EN/SET Off Timeout vs. Input Voltage


Turn-On Characteristic ( $\mathrm{I}_{\text {out } 1}=\mathrm{I}_{\text {out } 2}=20 \mathrm{~mA}$ )


Time ( $50 \mu \mathrm{~s} / \mathrm{div}$ )

EN/SET Latch Timeout vs. Input Voltage


Turn-On Characteristic
(lout1 $=\mathrm{l}_{\text {out2 }}=20 \mathrm{~mA}$ )


Time ( $50 \mu \mathrm{~s} / \mathrm{div}$ )

Turn-Off Characteristic
$\left(\mathrm{I}_{\text {OUT } 1}=\mathrm{I}_{\text {OUT } 2}=20 \mathrm{~mA}\right.$ )


## Typical Characteristics

$\underset{\left(\text { lout } 1=I_{\text {out } 2}=20 \mathrm{~mA}\right)}{\text { Transits }}$


Time ( $50 \mu \mathrm{~s} / \mathrm{div}$ )

Turn-Off Transient Characteristic
( ${ }_{\text {out } 1}=\mathrm{I}_{\text {out } 2}=20 \mathrm{~mA}$ )


Turn-On Transient Characteristic ( $\mathrm{l}_{\text {out } 1}=\mathrm{I}_{\text {out }}=20 \mathrm{~mA}$ )


Turn-On Fall Time vs. Temperature

$\mathrm{V}_{\mathrm{IL}}$ vs. Input Voltage


## Functional Block Diagram



## Functional Description

The AAT4295/97 is comprised of three or six low-side N-channel MOSFET load switches primarily targeted for general purpose applications where several load circuits need to be connected to a common ground and controlled from a single microcontroller GPIO output. When a given switch is enabled, the respective switch connects the load input (S1 to S3 for the AAT4295 and S1 to S6 for the AAT4297) to ground through the $N$-channel MOSFET. Each low-side N-channel MOSFET transistor has a typical on resistance ( $\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}$ ) of $2 \Omega$ when operating from a 3.6 V supply. Both the AAT4295 and AAT4297 have been designed to operate with an input voltage range of 1.8 V to 5.5 V , making them ideal for battery-powered applications.

These devices may be used for load switching applications such as RGB LED fashion lighting, display or keypad backlight LEDs, miscellaneous indicator LED lamps, as well as audio and RF circuits or any other system with a power requirement that does not exceed the thermal dissipation limits of the load switch and device package.
Each switch input may be represented by the following circuit (Figure 1) and simplified equivalent model (Figure $2)$.

The state of each switch is controlled via the EN/SET pin using Skyworks' S² ${ }^{2}$ wire interface. To enable a respective switch, a series of clocked pulses should be applied to the EN/SET pin. The number of pulses clocked will determine the switch configuration based on the truth table given in Table 1. At the end of the serial pulse data
set, the EN/SET set pin should be held high to latch the clocked data and enable the desired switch configuration. When the device is enabled with the EN/SET held to a logic high state, the quiescent current consumption will typically increase to $3 \mu \mathrm{~A}$ at normal ambient room temperatures. If output sequencing of the switches is not necessary, all of the switches may be turned on simultaneously on the first rising edge of the EN/SET pin by simply pulling the EN/SET to a logic high level. The default configuration for one clock pulse is to enable all switches to the "on" state. However, if output sequencing is desired, a series of pulses on the EN/SET pin will set the outputs to the desired state (refer to Table 2 for output settings). For LED lighting applications, the EN/ SET line may be clocked at rates up to 1 MHz , allowing the user to not only control brightness, but (in the case of color RGB LEDs) color as well.

## Output Settings

The ON/OFF state of the MOSFET switches is controlled by the EN/SET serial data input. An internal control counter is clocked on the rising edge of the EN/SET pin, and is decoded into the 8 possible states of the MOSFET for the AAT4295 (see Table 1) and 64 possible states for the AAT4297 (see Table 2). The counter rolls over after 8 clocks and the table repeats.

| Clock | OUT3 | OUT2 | OUT1 |
| :---: | :---: | :---: | :---: |
| 1 | on | on | on |
| 2 | on | on | off |
| 3 | on | off | on |
| 4 | on | off | off |
| 5 | off | on | on |
| 6 | off | on | off |
| 7 | off | off | on |
| 8 | off | off | off |

Table 1: AAT4295 EN/SET Settings.
The $S^{2}$ Cwire interface relies on the number of rising edges of the EN/SET pin to load the internal register to a desired count. $\mathrm{S}^{2}$ Cwire control latches data after the EN/SET pin has been held high for the preset latch enable time ( $\mathrm{T}_{\mathrm{LAT}}$ ). The interface records rising edges of the EN/SET pin and decodes them into one of 8 states for the AAT4295 or to one of 64 states for the AAT4297, as indicated in Tables 1 and 2. The counter can be clocked at speeds up to 1 MHz , so different switch combinations may be clocked in lighting applications without any visible perception to the user. Alternatively, the EN/SET clock pulses may be entered one at a time for the desired setting. The first rising edge of EN/SET enables the IC and turns all the switches ON. Once the final clock cycle is received, the EN/SET pin is held high to maintain the device setting. The device is disabled $500 \mu$ ( $T_{\text {OFF }}$ ) after the EN/SET pin transitions to a logic low state (see Figure 3).


Figure 2: Simplified Equivalent Model.

| Clock | AAT4297 |  |  |  |  |  | Clock | AAT4297 |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | OUT6 | OUT5 | OUT4 | AAT4295 (only) |  |  |  | OUT6 | OUT5 | OUT4 | OUT3 | OUT2 | OUT1 |
|  |  |  |  | OUT3 | OUT2 | OUT1 |  |  |  |  |  |  |  |
| 1 | on | on | on | on | on | on | 33 | off | on | on | on | on | on |
| 2 | on | on | on | on | on | off | 34 | off | on | on | on | on | off |
| 3 | on | on | on | on | off | on | 35 | off | on | on | on | off | on |
| 4 | on | on | on | on | off | off | 36 | off | on | on | on | off | off |
| 5 | on | on | on | off | on | on | 37 | off | on | on | off | on | on |
| 6 | on | on | on | off | on | off | 38 | off | on | on | off | on | off |
| 7 | on | on | on | off | off | on | 39 | off | on | on | off | off | on |
| 8 | on | on | on | off | off | off | 40 | off | on | on | off | off | off |
| 9 | on | on | off | on | on | on | 41 | off | on | off | on | on | on |
| 10 | on | on | off | on | on | off | 42 | off | on | off | on | on | off |
| 11 | on | on | off | on | off | on | 43 | off | on | off | on | off | on |
| 12 | on | on | off | on | off | off | 44 | off | on | off | on | off | off |
| 13 | on | on | off | off | on | on | 45 | off | on | off | off | on | on |
| 14 | on | on | off | off | on | off | 46 | off | on | off | off | on | off |
| 15 | on | on | off | off | off | on | 47 | off | on | off | off | off | on |
| 16 | on | on | off | off | off | off | 48 | off | on | off | off | off | off |
| 17 | on | off | on | on | on | on | 49 | off | off | on | on | on | on |
| 18 | on | off | on | on | on | off | 50 | off | off | on | on | on | off |
| 19 | on | off | on | on | off | on | 51 | off | off | on | on | off | on |
| 20 | on | off | on | on | off | off | 52 | off | off | on | on | off | off |
| 21 | on | off | on | off | on | on | 53 | off | off | on | off | on | on |
| 22 | on | off | on | off | on | off | 54 | off | off | on | off | on | off |
| 23 | on | off | on | off | off | on | 55 | off | off | on | off | off | on |
| 24 | on | off | on | off | off | off | 56 | off | off | on | off | off | off |
| 25 | on | off | off | on | on | on | 57 | off | off | off | on | on | on |
| 26 | on | off | off | on | on | off | 58 | off | off | off | on | on | off |
| 27 | on | off | off | on | off | on | 59 | off | off | off | on | off | on |
| 28 | on | off | off | on | off | off | 60 | off | off | off | on | off | off |
| 29 | on | off | off | off | on | on | 61 | off | off | off | off | on | on |
| 30 | on | off | off | off | on | off | 62 | off | off | off | off | on | off |
| 31 | on | off | off | off | off | on | 63 | off | off | off | off | off | on |
| 32 | on | off | off | off | off | off | 64 | off | off | off | off | off | off |

Table 2: Output Settings.


Figure 3: EN/SET Timing Diagram.

## Application Information

## External Component Selection

The AAT4295 and AAT4297 have been designed so that no external parts are required for the device to function as a general purpose three- or six-position low-side switch. For some applications, the addition of bypass capacitors or pull-up or pull-down resistors may be desired to improve overall system performance.

For lighting applications, such as controlling RGB LEDs, keypad or display backlight LEDs, or photo flash LEDs, no bypass capacitors are necessary. For other general purpose load switching applications which may use some or all of the outputs to switch light load current levels to application circuits, good engineering practice would dictate the use of small bypass capacitors placed on the $\mathrm{V}_{\mathrm{Cc}}$ input and each switch connection that is used to conduct current from the load to ground. The use of small ceramic capacitors between the input and output nodes will aid in reducing line and load transient response effects when the load switch on a given output is turned on or off.

## Input Capacitor

Typically, a $0.1 \mu \mathrm{~F}$ capacitor is recommended for $\mathrm{C}_{\mathrm{IN}}$ in most applications to provide input line transient response immunity to small changes in the input supply. $A C_{I N}$ capacitor is not required for basic operation. If used, $\mathrm{C}_{\mathrm{IN}}$ should be located as close to the device $\mathrm{V}_{\text {IN }}$ pin as practically possible.

There is no specific capacitor equivalent series resistance (ESR) requirement for $\mathrm{C}_{\mathrm{IN}}$; however, for higher current operation, ceramic capacitors are recommended for $\mathrm{C}_{\mathrm{IN}}$ due to their inherent capability over tantalum or aluminum electrolytic capacitors to withstand input current surges from low impedance sources, such as batteries in portable devices.

## Output Capacitor

For typical applications where the AAT4295/97 is used for LED lighting control, no output capacitors are required because the end load is not sensitive to device turn-on or turn-off transient effects.
For improved load transient response in systems using the AAT4295/97 for load switching, the addition of a small output capacitor placed between the output pins
and ground can have a beneficial effect. A $0.1 \mu \mathrm{~F}$ ceramic capacitor is suggested as a reasonable value for an output capacitor.

The output capacitor has no specific capacitor type or ESR requirement. If desired, $\mathrm{C}_{\text {оut }}$ may be increased to a value greater than $0.1 \mu \mathrm{~F}$ without limit to accommodate any load transient condition without adversely affecting the device turn-on slew rate time.

## Thermal Considerations

The AAT4295 and AAT4297 are designed to sink a continuous load current to ground when a respective switch is enabled via the S2Cwire control. The limiting characteristic for maximum safe operating load current through a given switch or set of switches is package power dissipation. In order to obtain high operating currents, careful device layout and circuit operating conditions must be taken into account. At any given ambient temperature $\left(T_{A}\right)$, the maximum package power dissipation can be determined by the following equation:

$$
P_{D(M A X)}=\frac{T_{J(M A X)}-T_{A}}{\theta_{J A}}
$$

Constants for the AAT4295 are maximum junction temperature, $\mathrm{T}_{J \text { (MAX) }}=125^{\circ} \mathrm{C}$, and package thermal resistance, $\theta_{\mathrm{JA}}=225^{\circ} \mathrm{C} / \mathrm{W}$. Worst case conditions are calculated at the maximum operating temperature, where $T_{A}$ $=85^{\circ} \mathrm{C}$. Typical conditions are calculated under normal ambient conditions, where $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$. At $25^{\circ} \mathrm{C}$ ambient, the AAT4295 is capable of dissipating 444.4 mW of power and the AAT4297 is capable of dissipating 625 mW of power. At $85^{\circ} \mathrm{C}$ ambient, the AAT4295 is capable of dissipating 177.8 mW of power and the AAT4297 can dissipate 250 mW .

The power dissipation of any given MOSFET switch is limited by its respective on resistance ( $R_{D S}$ ). The $R_{D S}$ of any given MOSFET switch is controlled by the applied gate voltage to the switch, which is set by the applied $\mathrm{V}_{\mathrm{Cc}}$ supply and the ambient operating temperature. Switch RDS for the AAT4295 or AAT4297 may be estimated by using the $R_{D s}$ versus Temperature curve in the Typical Characteristics section of this datasheet. The maximum current of any given switch can be calculated for a given operating temperature and $\mathrm{V}_{\mathrm{cc}}$ supply level. The corresponding $R_{D S}$ is determined by use of the $R_{D S}$ vs. Temperature curve for the given $\mathrm{V}_{\mathrm{cc}}$.

Given the maximum package power dissipation and operating temperature, the maximum current through any switch or combination of switches can be calculated using the following formula:

$$
\mathrm{I}_{\mathrm{SWITCH}(\mathrm{MAX})}=\left(\frac{\mathrm{P}_{\mathrm{D}(\mathrm{MAX})}}{\mathrm{R}_{\mathrm{DS}}}\right)^{\frac{1}{2}}
$$

Example: If all the switches on an AAT4295 were closed simultaneously, each switch could handle up to 271 mA of current at $25^{\circ} \mathrm{C}$ for a total of 813 mA . For the same set of operating conditions at $25^{\circ} \mathrm{C}$, the AAT4297 can handle up to 208 mA per switch for a total of 1.25 A for all six switches. If the load current for a desired application exceeds the recommended current at a given temperature, two or more switches may be operated in parallel as long as the overall power dissipation of the device package is not exceeded. If different current levels are passed through different switches on a given device, then one should total up the power dissipation for each switch and assure the sum of the power dissipation does not exceed the power rating for the package.

## Application Circuits

Today, many mobile phones and similar products contain RGB LED fashion lighting, LCD display and sub-display, as well as keypad backlighting and photo flash LEDs. Due to the nature of common anode RGB LEDs, the AAT4295 and AAT4297 make ideal low-cost lighting control solutions. In general, most types of LEDs can be controlled via a low-side MOSFET switch and current limiting ballast resistor. The following application circuits (Figures 4 through 7) show voltage boosting charge
pumps to power RGB and flash LEDs. However, if a voltage or current source is already available in a given product design, the charge pump circuit block may be replaced with the existing power source solution. Since both the AAT4295 and AAT4297 require only one GPIO line from the system microcontroller to enable and disable all the switches via the EN/SET input, these solutions can provide a simple way to add lighting solutions to existing design platforms.

## Driving LED Loads

When driving LEDs with a voltage source, series ballast resistors must be used to limit the LED forward current. The LED current will vary with supply voltage and LED forward voltage. Most types of LEDs have forward voltage specifications ranging from 2.0 V to 5.0 V . When controlling an LED of any type with a low-side MOSFET switch, the necessary series ballast resistor value can be calculated from the following formula:

$$
R_{\text {BALLAST }}=\frac{\left(\mathrm{V}_{\text {IN }}-\mathrm{V}_{\mathrm{F}}\right)}{\mathrm{I}_{\text {LED }}}-\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}
$$

Where:
$\mathrm{R}_{\text {BALLAST }}$ is the value of resistor to be placed in series with the LED ( $\Omega$ ).
$\mathrm{V}_{\text {IN }}$ is the input supply voltage to the device ( V ).
$V_{F}$ is the forward voltage of the LED (V).
$\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}$ is the resistance of the switch when it is turned on ( $\Omega$ ).
$\mathrm{I}_{\text {LED }}$ is the desired operating current of the LED (A).


Figure 4: Single RGB LED Fashion Light Solution Using an AAT4295.


Figure 5: Dual RGB LED Fashion Light Solution Using an AAT4297.


Figure 6: RGB LED Fashion Light With a Dual Mode Light/Strobe Flash LED Solution Using an AAT4297.


Figure 7: Total Lighting Control Solution Using an AAT4297. Includes RGB Fashion Light, Main Display and Sub-Display LCD Backlight, and Photo Flash LED.

## Ordering Information

| Package | Marking $^{1}$ | Part Number (Tape and Reel) $^{2}$ |
| :---: | :---: | :---: |
| SC70JW-8 | RBXYY | AAT4295IJS-T1 |
| TSOPJW-12 | RCXYY | AAT4297ITP-T1 |

Skyworks Green ${ }^{\text {TM }}$ products are compliant with all applicable legislation and are halogen-free. For additional information, refer to Skyworks Definition of Green ${ }^{T M}$, document number SQ04-0074.

## Package Information

## SC70JW-8



All dimensions in millimeters.

## TSOPJW-12



All dimensions in millimeters.

Copyright © 2012 Skyworks Solutions, Inc. All Rights Reserved.


 responsibility whatsoever for conflicts, incompatibilities, or other difficulties arising from any future changes.
 under, including the sale, distribution, reproduction or use of Skyworks products, information or materials, except as may be provided in Skyworks Terms and Conditions of Sale.



 THE USE OF THE MATERIALS OR INFORMATION, WHETHER OR NOT THE RECIPIENT OF MATERIALS HAS BEEN ADVISED OF THE POSSIBILITY OF SUCH DAMAGE.

 use or sale.

 design, or damage to any equipment resulting from the use of Skyworks products outside of stated published specifications or parameters.
 identification purposes only, and are the property of their respective owners. Additional information, including relevant terms and conditions, posted at www.skyworksinc.com, are incorporated by reference.


[^0]:     specified is not implied. Only one Absolute Maximum Rating should be applied at any one time.
    2. Mounted on an FR4 board.
    3. Derated $4.4 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $25^{\circ} \mathrm{C}$.
    4. Derated $6.25 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $25^{\circ} \mathrm{C}$.

[^1]:     tion with statistical process controls.
    2. $T_{O N}$ is the time after latch timeout to $10 \%$ of the output voltage. See Figure 1, Timing Diagram.

