

LT1612

#### Synchronous, Step-Down 800kHz PWM DC/DC Converter

#### FEATURES

- Operates from Input Voltage As Low As 2V
- Internal 0.7A Synchronous Switches
- Uses Ceramic Input and Output Capacitors
- 620mV Reference Voltage
- 800kHz Fixed Frequency Switching
- Programmable Burst Mode Operation
- Low Quiescent Current: 160µA
- 8-Lead MSOP or SO Package

#### **APPLICATIONS**

- Portable Devices
- Lithium-Ion Step-Down Converters
- 5V to 3.3V Conversion
- 2-Cell Alkaline Step-Down Converters

#### DESCRIPTION

The LT<sup>®</sup>1612 is an 800kHz, synchronous step-down DC/ DC converter that operates from an input voltage as low as 2V. Internal  $0.45\Omega$  switches deliver output currents up to 500mA, and the 800kHz switching frequency allows the use of small, low value ceramic input and output capacitors. Input voltage ranges from 5.5V down to 2V and output voltage can be set as low as the 620mV reference. The device features Burst Mode<sup>TM</sup> operation, keeping efficiency high at light loads. Burst Mode operation can be defeated by pulling the MODE pin high, enabling constant switching throughout the load range for low noise.

No-load quiescent current is  $160\mu$ A and shutdown current is less than  $1\mu$ A. The device is available in 8-lead SO and MSOP packages.

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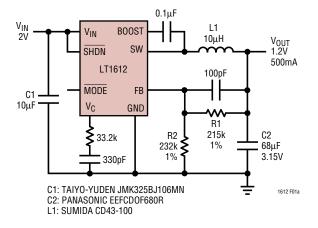
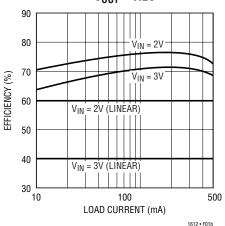


Figure 1. 2V to 1.2V Converter

### Efficiency for LT1612 vs Linear Regulator $V_{OUT} = 1.2V$



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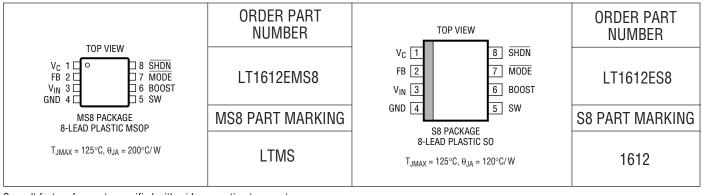
#### TYPICAL APPLICATION

#### ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage (V <sub>IN</sub> )	5.5V
SW Pin Voltage	
FB Pin Voltage	
V <sub>C</sub> Pin Voltage	2V
SHDN Pin Voltage	
MODE Pin Voltage	

BOOST Pin Voltage	V <sub>IN</sub> + 5.5V
Junction Temperature	125°C
Operating Temperature Range (Note 2)40	°C to 85°C
Storage Temperature Range65°	C to 150°C
Lead Temperature (Soldering, 10 sec)	300°C

### PACKAGE/ORDER INFORMATION



Consult factory for parts specified with wider operating temperature ranges.

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes specifications which apply over the full operating temperature range, otherwise specifications are T<sub>A</sub> = 25°C, V<sub>IN</sub> = V<sub>SHDN</sub> = 3V

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
Ι <sub>Q</sub>	Quiescent Current	MODE = 5V			1	2	mA
		MODE = 0V, Not Switching	•		160	220	μA
		SHDN = 0V	•			1	μA
V <sub>FB</sub>	FB Voltage			0.605	0.62	0.635	V
			•	0.60	0.62	0.635	V
	FB Line Regulation		•		0.02	0.15	%/V
	FB Pin Bias Current (Note 3)		•		7	50	nA
g <sub>m</sub>	Error Amplifier Transconductance				250		µmhos
	Min Input Voltage			2			V
	Max Input Voltage					5.5	V
f <sub>OSC</sub>	Oscillator Frequency			700	800	900	kHz
			•	550		1100	kHz
	f <sub>OSC</sub> Line Regulation				1		%/V
	Maximum Duty Cycle			85	90		%
			•	80			%
	Shutdown Threshold	Minimum Voltage for Active		2			V
		Maximum Voltage for Shutdown	•			0.2	V

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# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes specifications which apply over the full operating temperature range, otherwise specifications are $T_A = 25^{\circ}C$ , $V_{IN} = V_{SHDN} = 3V$

SYMBOL	PARAMETER	CONDITIONS			MIN	ТҮР	MAX	UNITS
	SHDN Pin Current	$\overline{SHDN} = 2V$				10	15	μA
		SHDN = 5V			30	45	μA	
	BOOST Pin Current	BOOST = V <sub>IN</sub> + 2V			4		mA	
	Switch Current Limit (Note 4)	Duty Cycle = 0%	MODE = OV		600	710	900	mA
			MODE = 5V		550	650	900	mA
	Burst Mode Operation Current Limit	MODE = 0V				180		mA
	Switch Voltage Drop	I <sub>SW</sub> = 500mA				200	280	mV
	Rectifier Voltage Drop	I <sub>RECT</sub> = 500mA				300	400	mV
	SW Pin Leakage	$V_{SW} = 5V, V_{\overline{SHDN}} = 0V$		•			1	μA

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

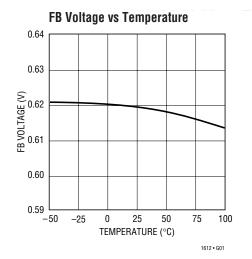
Note 2: The LT1612E is guaranteed to meet performance specifications from 0°C to 70°C. Specifications over the -40°C to 85°C operating

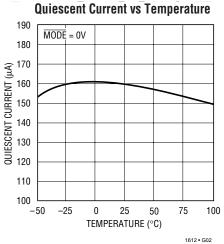
temperature range are assured by design, characterization and correlation with statistical process controls.

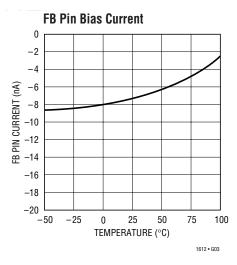
Note 3: Bias current flows out of the FB pin.

Note 4: Duty cycle affects current limit due to slope compensation.

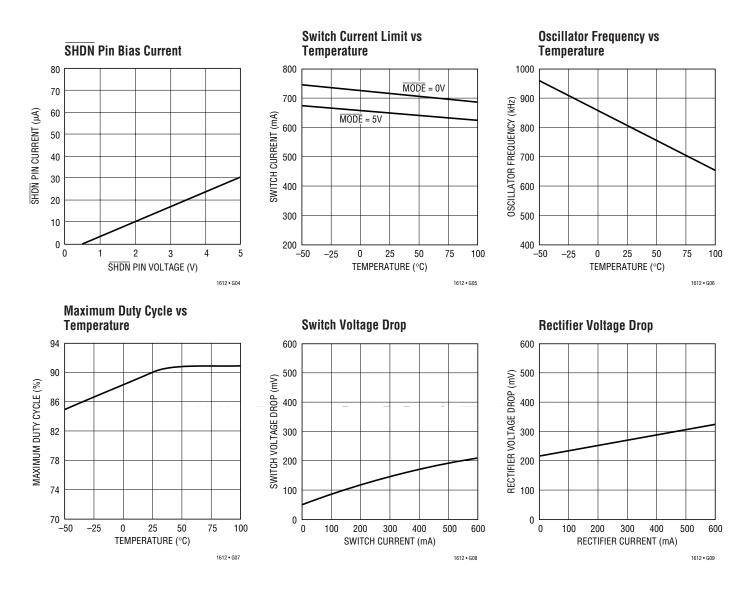
### **TYPICAL PERFORMANCE CHARACTERISTICS**







### TYPICAL PERFORMANCE CHARACTERISTICS



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#### PIN FUNCTIONS

 $V_C$  (Pin 1): Compensation Pin. This is the current sink/ source output of the error amplifier. By connecting an RC network from this pin to ground, frequency response can be tuned for a wide range of circuit configurations. The voltage at this pin also sets the current limit, and if grounded, the switch will remain in the OFF state.

**FB (Pin 2):** Feedback Pin. This pin is the negative input to the error amplifier. Connect the resistor divider tap to this point which sets  $V_{OUT}$  according to:

 $V_{OUT} = 0.62V (1 + R1/R2)$ 

 $V_{IN}$  (Pin 3): Supply Pin. Bypass capacitor C1 must be right next to this pin.

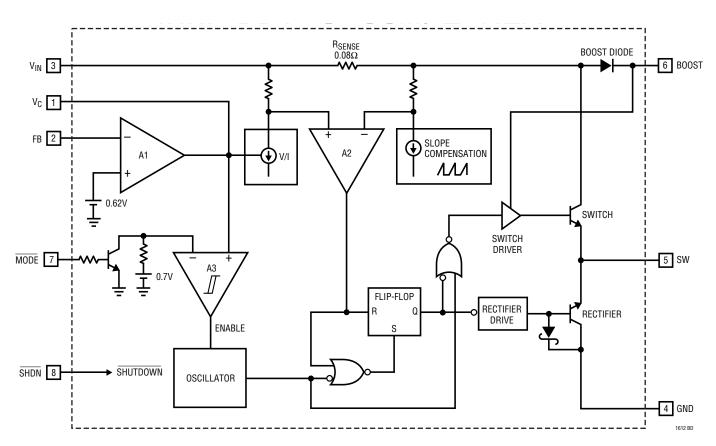
**GND (Pin 4):** Ground Pin. Connect directly to local ground plane.

**SW (Pin 5):** Switch Pin. Connect inductor and boost capacitor here. Minimize trace area at this pin to keep EMI down.

**BOOST (Pin 6):** This is the supply pin for the switch driver and must be above  $V_{\text{IN}}$  by 1.5V for proper switch operation. Connect the boost capacitor to this pin.

**MODE** (Pin 7): Burst Mode Operation Disable Pin. For continuous switching operation (low noise), pull this pin above 2V. For Burst Mode operation which gives better light load efficiency, tie to ground. Output ripple voltage in Burst Mode operation is typically 30mV<sub>P-P</sub>. See applications section for more information about this function.

**SHDN** (Pin 8): Shutdown Pin. Pull this pin low for shutdown mode. Tie to a voltage between 2V and 5.5V for normal operation.



### **BLOCK DIAGRAM**

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### OPERATION

The LT1612 employs fixed frequency, current mode control. This type of control uses two feedback loops. The main control loop sets output voltage and operates as follows: A load step causes  $V_{OUT}$  and the FB voltage to be perturbed slightly. The error amplifier responds to this change in FB by driving the  $V_C$  pin either higher or lower. Because switch current is proportional to the  $V_C$  pin voltage, this change causes the switch current to be adjusted until  $V_{OUT}$  is once again satisfied. Loop compensation is taken care of by an RC network from the  $V_C$  pin to ground.

Inside this main loop is another that sets current limit on a cycle-by-cycle basis. This loop utilizes current comparator A2 to control peak current. The oscillator runs at 800kHz and issues a set pulse to the flip-flop at the beginning of each cycle, turning the switch on. With the switch now in the ON state the SW pin is effectively connected to  $V_{IN}$ . Current ramps up in the inductor linearly at a rate of  $(V_{IN} - V_{OUT})/L$ . Switch current is set by the  $V_C$  pin voltage and when the voltage across  $R_{SENSE}$  trips the current comparator, a reset pulse will be generated and the switch will be turned off. Since the inductor is now loaded up with current, the SW pin will fly low and trigger the rectifier to turn on. Current will flow through the rectifier decreasing at a rate of  $V_{OUT}/L$  until the oscillator issues a new set pulse, causing the cycle to repeat.

If the load is light and  $V_C$  decreases below A3's trip point, the device will enter the Burst Mode operation region (the MODE pin must be at ground or floating). In this state the oscillator and all other circuitry except the reference and comparator A3 are switched on and off at low frequency. This mode of operation increases efficiency at light loads but introduces low frequency voltage ripple at the output. For continuous switching and no low frequency output voltage ripple, pull the MODE pin high. This will disable comparator A3 which forces the oscillator to run continuously.

#### **Layout Hints**

The LT1612 switches current at high speed, mandating careful attention to layout for proper performance. *You will not get advertised performance with careless layout.* Figure 2 shows recommended component placement for a buck (step-down) converter. Follow this closely in your PC layout. Note the direct path of the switching loops. Input capacitor C1 must be placed close (<5mm) to the IC package. As little as 10mm of wire or PC trace from C<sub>IN</sub> to V<sub>IN</sub> will cause problems such as inability to regulate or oscillation.

The ground terminal of input capacitor C1 should tie close to Pin 4 of the LT1612. Doing this reduces dl/dt in the ground copper which keeps high frequency spikes to a minimum. The DC/DC converter ground should tie to the PC board ground plane at one place only, to avoid introducing dl/dt in the ground plane.

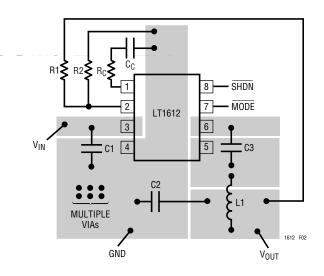


Figure 2. Recommended Component Placement. Traces Carrying High Current are Direct. Trace Area at FB Pin and V<sub>C</sub> Pin Is Kept Low. Lead Length to Battery Should Be Kept Short



#### OPERATION

#### **Burst Mode Operation Defeat**

To maintain high efficiency at light loads, the LT1612 will automatically shift into Burst Mode operation (MODE = 0V or floating). In this mode of operation the oscillator and switch drive circuitry is alternately turned on and off, reducing quiescent current to 160 $\mu$ A. This reduces power

consumption but also adds low frequency voltage ripple to the output. Figure 3 shows switching waveforms for a 5V to 3.3V converter running in Burst Mode operation. Output voltage ripple is approximately  $20mV_{P-P}$ . If the MODE pin is pulled high, Burst Mode operation will be inhibited and the oscillator runs continuously with no low frequency ripple at the output. See Figures 4 and 5.

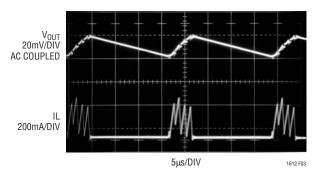


Figure 3. Output Voltage Ripple is  $20mV_{P\mbox{-}P}$  for the Circuit of Figure 1

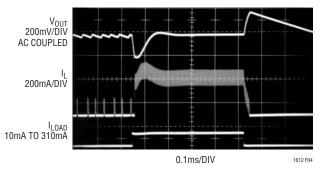


Figure 4. <u>Transient Response for the Circuit of Figure 1</u> with the MODE Pin Tied to Ground or Floating

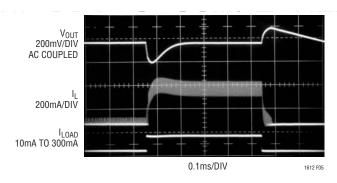
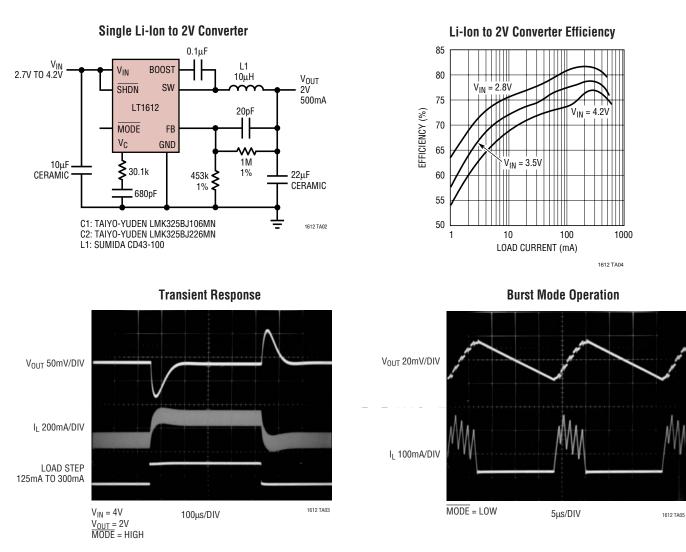
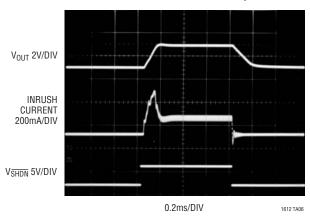


Figure 5. With the MODE Pin Tied High, Low Frequency Output Voltage Ripple Is No Longer Present



### TYPICAL APPLICATIONS



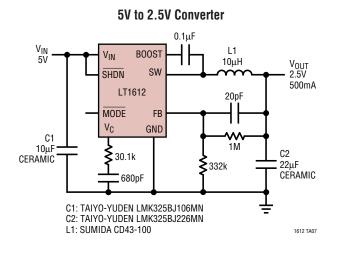


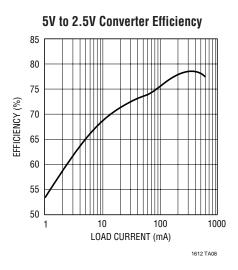
Inrush Current at Start-Up

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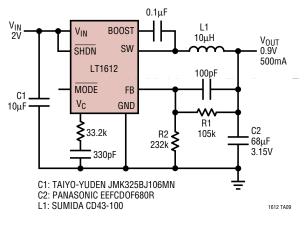


#### TYPICAL APPLICATIONS

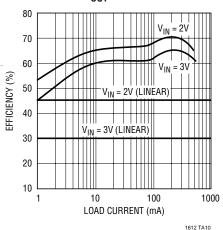




2V to 0.9V Converter



Efficiency for LT1612 vs Linear Regulator.  $V_{OUT}$  = 0.9V.





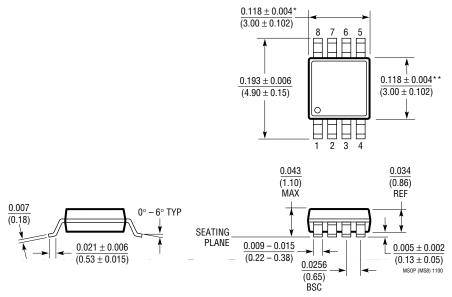
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#### **PACKAGE DESCRIPTION** Dimension in inches (millimeters) unless otherwise noted.

MS8 Package 8-Lead Plastic MSOP (LTC DWG # 05-08-1660)



\* DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

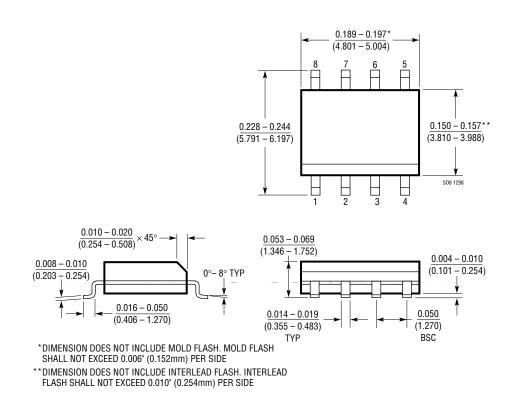
\*\* DIMENSION DOES NOT INCLUDE INTERLEAD FLASED OR PROTRUSIONS. INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

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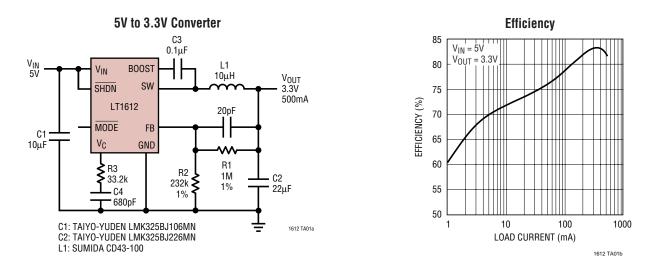
#### **PACKAGE DESCRIPTION** Dimension in inches (millimeters) unless otherwise noted.

S8 Package 8-Lead Plastic Small Outline (Narrow 0.150) (LTC DWG # 05-08-1610)





### TYPICAL APPLICATION



### **RELATED PARTS**

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PART NUMBER	DESCRIPTION	COMMENTS
LTC®1474	Low I <sub>Q</sub> Step-Down Switching Regulator	$10\mu A~I_Q,~V_{IN}$ from 3V to 18V, MSOP Package up to 300mA
LT1616	600mA, 1.4MHz Step-Down Regulator in SOT-23	V <sub>IN</sub> from 3.6V to 25V, SOT-23 Package
LTC1701	SOT-23 Step-Down Switching Regulator	500mA in SOT-23 Package, 1MHz Switching Frequency
LTC1707	Monolithic Synchronous Step-Down Switching Regulator	500mA, V <sub>IN</sub> from 2.65V to 8.5V
LTC1772	Constant Frequency Step-Down Controller in SOT-23	High Current, High Efficiency: Up to 94%
LTC1877/LTC1878	High Efficiency, Monolithic Synchronous Step-Down Regulator	10µA I <sub>Q</sub> , 2.65≤ V <sub>IN</sub> ≤ 10V, MSOP Package up to 600mA
LTC3404	1.4MHz High Efficiency Monolithic Synchronous Step-Down Reg	10μA I <sub>Q</sub> , High Efficiency: up to 95%, MSOP Package



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